

# Developments for collective Thomson scattering equipment with a sub-THz gyrotron in LHD

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**Abstract.** Plan of collective Thomson scattering (CTS) experiment for the Large Helical Device (LHD) in NIFS with a 303 GHz gyrotron is under way. Use of a sub-THz gyrotron expands the CTS-applicable region of plasma parameters. In LHD, sub-THz CTS can be applied to the high density operation region, plasmas with impurity hole, etc. Moreover, sub-THz CTS is expected to be free from ECE noise. Its “collective” use with 77 GHz and 154 GHz CTS will compose a powerful diagnostic system. A high power sub-THz gyrotron with a frequency of 303 GHz has been developed. Its maximum power is 320 kW. It oscillates in pulse mode and the maximum pulse width is around 100  $\mu$ s, which is sufficient for use in CTS experiments. A whispering gallery mode TE<sub>22,2</sub> was adopted for this gyrotron to avoid mode competition. Careful frequency measurement has proved purely single mode oscillation of the TE<sub>22,2</sub> mode including turn-on and turn-off phases of the oscillation pulse. This is consistent with mode competition calculations taking account of a finite voltage rise time. A low loss transmission line is necessary for CTS. We have two possibilities. One is a new line with 1.25 inch corrugated waveguides that are optimized for the 300 GHz band. Transmission test with the 303 GHz gyrotron has been carried out and a sufficiently low loss coefficient has been confirmed. The other is to use an existing line with 3.5 inch corrugated waveguides for lower frequencies such as 77 GHz and 154 GHz. Transmission test has been carried out with the 303 GHz gyrotron and a sufficiently low loss coefficient has been confirmed also for 3.5 inch corrugated waveguides. An existing line with 3.5 inch corrugated waveguides will be used in the initial phase of 303 GHz CTS experiment.

## 1 Introduction

CTS diagnostics using gyrotrons for electron heating around 100 GHz have been developed [1, 2]. Sub-THz waves are promising for CTS because they do not suffer from refraction and absorption in plasmas and noise due to strong electron cyclotron emissions (ECEs). The power source for CTS should be more than 100 kW to realize a suitable signal-to-noise (SN) ratio [3]. Then, a high power sub-THz gyrotron has been developed for use in CTS diagnostics in LHD.

A second harmonic gyrotron was initially developed. It realized single mode oscillation approaching 100 kW at 389 GHz. However, a competing fundamental harmonic mode prevented further increase in power [4]. We subsequently developed a fundamental harmonic prototype gyrotron with a whispering gallery mode (WGM) and used it to verify the design concept for stable single-mode oscillation [5]. The same design concept to use WGM mode was adopted in the design of a practical gyrotron and fabricated for use in CTS diagnostics in the LHD.

The designed operating frequency of the practical

gyrotron is 303.3 GHz, which corresponds to the minimum for the ECE for standard LHD operation [6]. The cutoff density corresponding to 303 GHz is about  $10^{21}$  m<sup>-3</sup>. This gyrotron can be used in the high density operation mode of LHD.

The oscillation mode is the TE<sub>22,2</sub> mode, which is a moderately high order WGM similar to the TE<sub>14,2</sub> mode used in the prototype gyrotron. WGM mode is suitable to satisfy requirements for mode competition avoidance and low cavity surface ohmic loss simultaneously [5]. WGM modes were used in the initial stage of the development of gyrotrons for fusion plasma heating and stable oscillations without mode competition were shown [7, 8]. Volume modes were then used to reduce the ohmic loss for the continuous wave (cw) 1 MW class gyrotrons required for the International Thermonuclear Experimental Reactor (ITER) [9]. Because cw oscillation is not required in CTS diagnostics, the 303 GHz gyrotron is operated in pulse mode. WGM modes are promising for complete suppression of spurious modes, including the turn-on and turn-off phases of each oscillation pulse.

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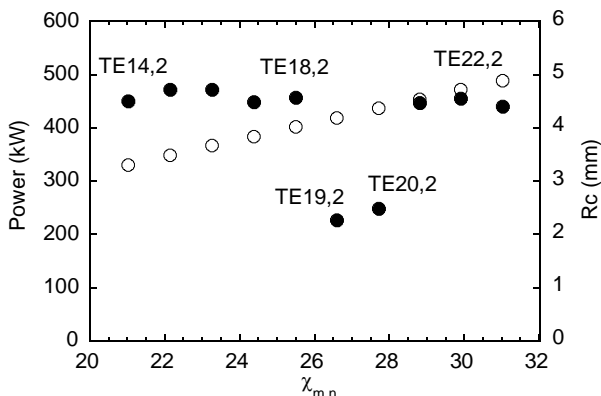
A low loss transmission line is necessary for realization of CTS diagnostics because the gyrotron is located far away from LHD. The transmission line will be tens of meters long. A corrugated waveguide (CW) system is a prospective candidate of a low loss transmission line. Two types of CW are now considered. One is a new line with 1.25 inch CW optimized for 300 GHz band. The other is a 3.5 inch CW line already installed in LHD for electron heating with lower frequency gyrotrons.

We have first evaluated the transmission efficiency of the 1.25 inch CW with the 303 GHz gyrotron power. Five 2.085 m long waveguides were prepared for this test [10, 11]. An attenuation coefficient of 0.02 – 0.04 dB/m sufficiently small for use in CTS was obtained. However, theoretically calculated attenuation coefficients are smaller than 0.001 dB/m [12, 13]. Experimental errors in the measurement of the very small attenuation might cause the discrepancy. In addition, absorption by water vapor of the order of 0.01 dB/m should also be considered.

Since installation of a new line with the 1.25 inch CW is difficult for the time being, an existing line with 3.5 inch CW will be used in the initial stage of 303 GHz CTS. Then, we have measured the transmission efficiency with spare waveguides.

## 2 Design, fabrication and test of 303 GHz gyrotron

It was necessary to use WGM higher than that of the prototype gyrotron to extend the pulse width and the duty ratio up to 10%. Preliminary design consideration was carried out to find candidate modes of the practical gyrotron. Figure 1 plots the stationary oscillation power obtained with a gyrotron design code [14] including the competition with neighboring modes. This figure indicates that WGM modes are not always appropriate for a high power oscillation. The TE<sub>19,2</sub> and TE<sub>20,2</sub> modes subject to competition with neighboring modes. The TE<sub>22,2</sub> mode was finally adopted for the practical gyrotron [15]. The same mode was also employed in [16



**Fig. 1.** Power calculation of candidate modes (solid circles). The beam voltage and beam current were set at 65 kV and 20 A, respectively. Possible competing modes are included in the calculation. The cavity radius corresponding to the oscillation frequency of 303.3 GHz is also shown by open circles.

- 18]. Oscillation characteristics of the TE<sub>22,2</sub> mode were examined in detail with the gyrotron design code and stable single mode oscillation was expected. A power of more than 300 kW was predicted for a beam voltage  $V_k$  of 65 kV, a beam current  $I_B$  of 15 A, and a velocity pitch factor  $\alpha$  of 1.2. Ohmic loss on the cavity surface is safely smaller than 2 kW/cm<sup>2</sup> up to the duty ratio of 10%.

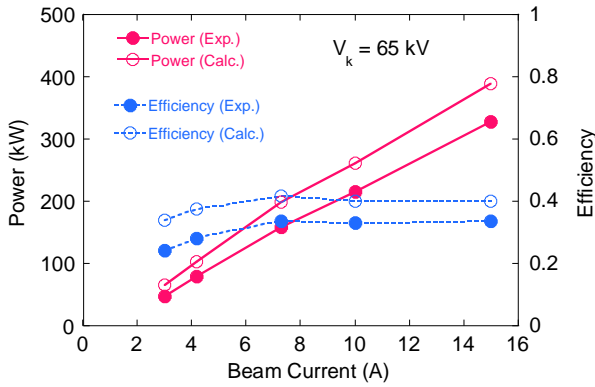
An electron gun was newly designed and optimized for the TE<sub>22,2</sub> mode by using the same design principle as reported in Ref. [19]. The design predicted electron beams of high quality with the expected  $\alpha$  values and sufficiently small velocity spreads up to  $I_B = 20$  A [20].



**Fig. 2.** Picture of the fabricated practical gyrotron. A Gaussian beam is radiated through the vacuum window at the left side.

Figure 2 is a picture of the fabricated gyrotron. This gyrotron is mounted on a liquid He-free 12 T SC magnet. The diameter of the room temperature bore of SC is 100 mm. An internal mode converter composed of a helical-cut Vlasov type launcher and four mirrors is installed into a rather narrow room. The vacuum window is made of a c-axis cut crystal sapphire disk. Its thickness is optimized for 303.3 GHz.

Oscillation tests have been carried out and a maximum power higher than 320 kW for  $V_k = 65$  kV and  $I_B = 15$  A was obtained [20, 21]. The oscillation efficiency was higher than 30%. Figure 3 shows the experimentally obtained power and efficiency as functions of the beam current. The calculated values are also plotted. In the calculation,  $\alpha$  values were evaluated for each experimental parameters with the E-GUN code. The experimental results are essentially consistent with calculations with  $\alpha$  ranging from 1.1 to 1.3, although the experimentally obtained powers are slightly lower than the calculated values. In this calculation instant voltage rise is assumed and interaction between the operating



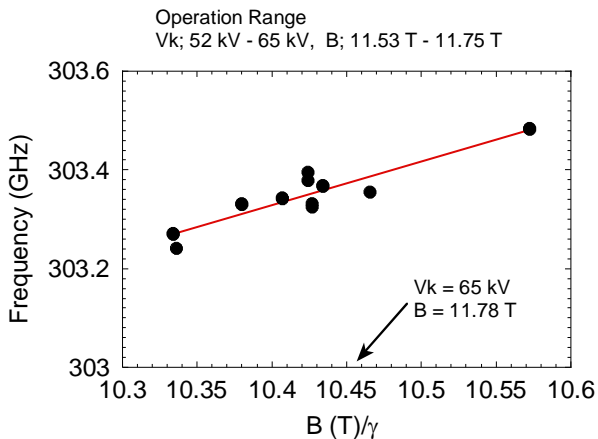
**Fig. 3.** Experimental and calculated values of the power and the efficiency are plotted. The TE<sub>21,2</sub> mode is included in the calculation of the power and efficiency.

TE<sub>22,2</sub> mode and the TE<sub>21,2</sub> mode is included. Although many other modes exist in between the two modes, interaction between the two modes is particularly strong.

The radiation pattern was measured using an infrared camera as the increasing temperature profile on a polyvinyl chloride (PVC) plate located in front of the vacuum window. A Gaussian like radiation pattern similar to that expected from the mode converter design calculation was obtained. The radiation pattern was somewhat elliptic and a small fraction of higher order modes was observed [20].

Measurement of the oscillation frequency using a Fabry-Perot interferometer verified single-mode oscillation of the TE<sub>22,2</sub> mode. The oscillation frequency was measured more accurately using a heterodyne receiver system. The measured frequency was 303.3 GHz when  $V_k = 62$  kV and the magnetic field at the cavity  $B = 11.59$  T. This frequency is equal to the designed TE<sub>22,2</sub> mode frequency to within an accuracy of the order of 10 MHz.

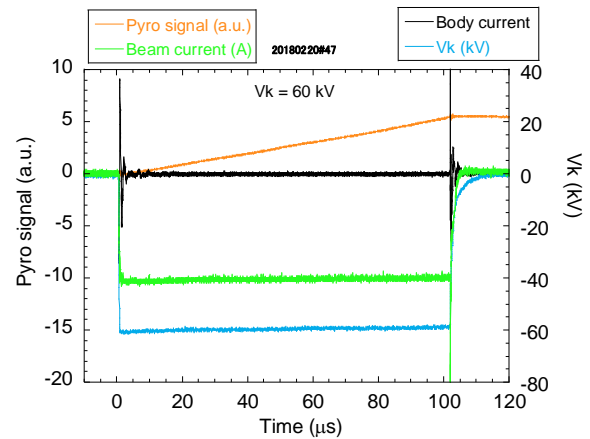
Frequency control is very important for application to CTS because a narrow band notch filter is used. A notch filter is necessary to protect the highly sensitive receiver system from the gyrotron-frequency wave stemming from scattering off the vacuum vessel. The oscillation



**Fig. 4.** Frequency variation as a function of a ratio of the magnetic field strength at the cavity  $B$  and the relativistic factor of the electron beam  $\gamma$ .

frequency slightly varies with the operation parameters such as  $V_k$  and  $B$ . We can control the oscillation frequency at around 303.3 GHz by appropriately changing  $V_k$  and  $B$ . Figure 4 indicates controllability of the frequency. The frequency variation with  $B$  and  $V_k$  was less than  $\pm 100$  MHz. The center frequency of the notch filter is 303.3 GHz and the full notch width is 0.5 GHz. The observed frequency variation around 303.3 GHz is safely smaller than the rejection width. The expected frequency width of the scattered signal is much wider than the notch width [3].

The pulse width was extended up to 100  $\mu$ s. Relevant waveforms are plotted in Fig. 5. The output power is approximately 200 kW for this case. The oscillation signal was measured using a pyroelectric detector in the time integration mode. The linear increase in this signal



**Fig. 5.** Waveforms of an 100  $\mu$ s oscillation. The beam voltage and beam current were set at 60 kV and 10 A, respectively.

indicates very stable power through the pulse width. This pulse width is sufficient for use in CTS diagnostics.

The frequency spectrum measured with the heterodyne receiver system was very narrow and stable across the pulse width. The frequency variation in one pulse is smaller than 10 MHz and the pulse to pulse difference is less than 20 MHz [15]. Parasitic modes were searched for over a wide frequency band, ranging from 290 GHz to 312 GHz, using the same heterodyne receiver system. No oscillation other than the TE<sub>22,2</sub> mode was observed throughout the pulse width, including during the turn-on and the turn-off phases. The dynamic range of measurement of about 40 dB is larger than the power range of the expected scattered signal [3].

The voltage of the modulation anode of the triode electron gun was set with a simple resistive divider. Stable and no-parasitic-mode oscillation without sophisticated voltage control of the modulation anode is likely due to a very fast rise time of the beam voltage. The beam voltage attains to 65 kV in about 0.25  $\mu$ s. A fast semiconductor switch realizes this short rise time. Suppression of parasitic modes for a voltage rise time shorter than 1  $\mu$ s was predicted [22]. A theoretical study just corresponding to the present gyrotron also predicts single mode oscillation of the TE<sub>22,2</sub> mode [23]. The frequency mismatch of the TE<sub>22,2</sub> mode is zero or even negative when the beam voltage starts to increase.

Analysis with a non-stationary axial field structure shows that a backward wave of the same transverse mode as the design mode is first soft-self excited and gradually transits to the high power gyrotron mode. This was predicted in [24].

### 3 Transmission test of 3.5 inch corrugated waveguides [25]

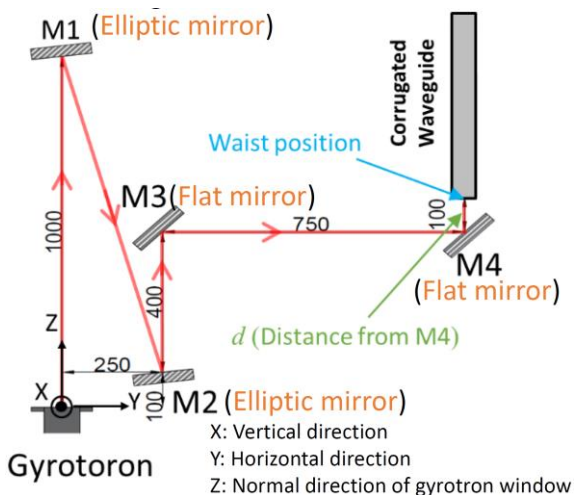
For CTS in LHD, it is necessary to transmit the gyrotron power to LHD with a low loss transmission line composed of corrugated waveguides over about 60 m. At the first stage of the 300 GHz CTS experiment, an existing 3.5 inch CW line is used for power transmission. Since this waveguide is not optimized for 300 GHz but for lower frequency gyrotrons for electron heating. Thus, transmission test is necessary. The corrugation parameters of the 3.5 inch CW are shown in Table I. The Bragg frequency of this waveguide is 187 GHz.

**Table 1.** Corrugation parameters of 3.5 in waveguide.

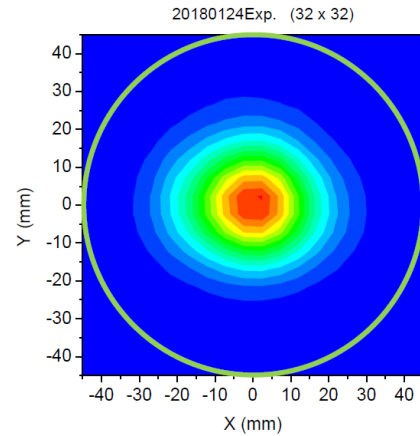
Waveguide diameter : a	88.9 mm
Corrugation depth: d	0.60 mm
Corrugation width: w	0.60 mm
Corrugation Pitch: p	0.80 mm

The output beam from the gyrotron is not axisymmetric and expands along propagation. Using a Matching Optics Unit (MOU), the output beam from the gyrotron was shaped into a Gaussian beam that couples to the HE<sub>11</sub> mode of the waveguide with a high coupling coefficient [26]. Figure 6 indicates setup of MOU. It is composed of two elliptic mirrors to tailor the beam radiated from the gyrotron to a Gaussian beam with the waist at the waveguide entrance position and two plane mirrors to adjust the beam direction.

The beam shape near the design waist position was



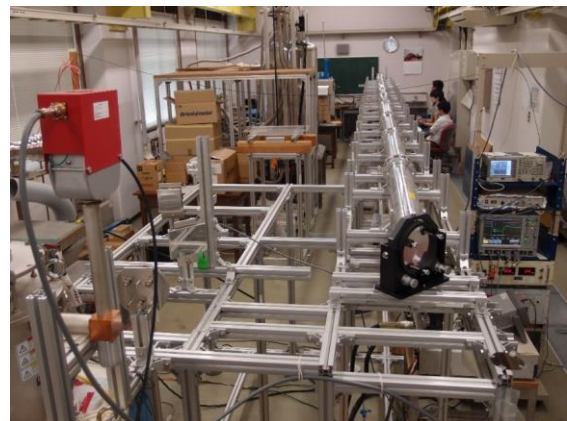
**Fig. 6.** Setup of Matching Optics Unit (MOU).



**Fig. 7.** Beam shape at the entrance position of 3.5 corrugated waveguide.

slightly elliptical. This is possibly because the gyrotron beam contains higher order modes. Figure 7 represents the beam shape at the waveguide entrance position measured as the distribution of temperature increase on a PVC plate. The green circle represents the waveguide cross section.

Figure 8 is a picture of the transmission test. Ten one-meter long waveguides were connected straight on a tight frame. The relative transmitted power was



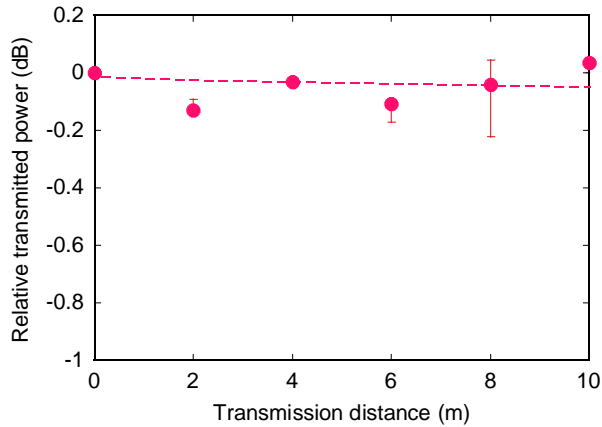
**Fig. 8.** Picture of transmission test.

evaluated at every 2 m distance by integrating the distribution of the temperature increase on a PVC plate over the waveguide cross section. The PVC plate was located just at the out put position for each distance measurement.

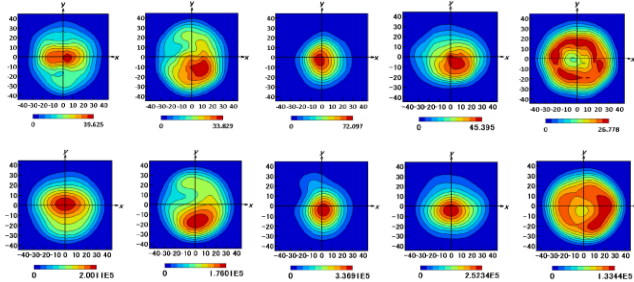
Figure 9 plots the relative power as a function of the transmission distance along 3.5 inch CW. Although measurements includes some error, attenuation is very low probably smaller than 0.01 dB/m. Correspondingly, the transmission loss is smaller than 0.6 dB for 60 m.

Figure 9 represents the total power transmitted to each measurement position. It includes attenuation of higher order modes excited in waveguides. In fact, owing to excitation of higher order modes, very complicated structures were observed at each measurement point.

Then, nonlinear mode content analysis was carried out with the same method as shown in [27]. In this



**Fig. 9.** Total power transmitted to each measurement position.



**Fig. 10.** Comparison between observed pattern of power distribution at each measured position (upper row) and the composed power distribution with calculated power fractions and phases (lower row).

analysis, 19 higher order modes were taken into account. Figure 10 represents the result of the analysis. Calculated power distributions as the composition of the  $HE_{11}$  mode and higher order modes well reproduces the measured power distribution. This composition takes account of relative phases between modes. The power fraction of  $HE_{11}$  mode was about 82%. Main higher order modes were co polarizing  $HE_{21}$  mode and  $HE_{12}$  mode with power fractions around 4%.

The transmission test shows that the existing 3.5 inch waveguide line can be used for 300 GHz CTS experiment. In the CTS experiment, more axisymmetric beam will be coupled into the waveguide.

## 4 Summary

Sub THz CTS experiment is under preparation with a pulse gyrotron in the 300 GHz band with collaboration between FIR UF and NIFS.

The design concept of the practical gyrotron is the same as that used in the prototype gyrotron. Higher mode  $TE_{22,2}$  was adopted to reduce Ohmic loss on the cavity surface. Stable single mode oscillation has been confirmed and the oscillation power and frequency are very stable over the pulse width. No parasitic mode was observed during the turn-on and turn-off phases. This is probably due to a finite voltage rise time. This gyrotron can be used in CTS diagnostics in LHD.

For power Transmission, an existing transmission line with 3.5 inch corrugated waveguides is used at the

first stage. Although this waveguide is not optimized for 300 GHz band, transmission test with the gyrotron have shown very low transmission loss.

Sub THz CTS experiment will start soon. Its “collective” use with 77 GHz and 154 GHz CTS will compose a powerful diagnostic system in LHD.

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